

## **Research Article**

### **Scrambled Audio Frequency Signal Transmission in a 3-slot STBC Scheme Based SC-FDMA Wireless Communication System**

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**Abstract:** Space time block coding (STBC), an effective and efficient transmit diversity scheme implemented Multi-input multi-output (MIMO) wireless communication systems have well been accepted as robust and efficient to combat detrimental effects of wireless fading channels. In comparison with STBC, a 3- slot STBC scheme (Orthogonal STBC) shows superior performance with achievement of rate-one full-diversity and reduced decoding complexity. In this paper, an effort has been made to study the performance of Orthogonal STBC based Single Carrier Frequency Division Multiple Access (SC-FDMA) based wireless communication system on scrambled audio signal transmission. The SC-FDMA system incorporates two channel coding (CRC and  $\frac{1}{2}$ -rated Convolutional), two linear channel equalization Successive Interference Cancellation (SIC) based Minimum Mean Square Error (MMSE-SIC) and Zero- Forcing (ZF-SIC) under QPSK, DQPSK and QAM digital modulations schemes. It is noticeable from simulation results that the system outperforms in retrieving transmitted audio signal under CRC, ZF-SIC and QAM schemes.

**Keywords:** SC-FDMA, 3-slot STBC, Channel coding, Linear channel equalization technique, Bit Error rate (BER), AWGN and Rayleigh fading channels.

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#### **INTRODUCTION**

Mobile communications has become an everyday commodity. Explosive demands for mobile data are driving changes in how mobile operators will need to respond to the challenging requirements of higher capacity and improved quality of user experience. Mobile communication technologies are often divided into generations, with 1G being the analog mobile radio systems of the 1980s, 2G the first digital mobile systems and 3G, the first mobile system handling broadband data. The Long-Term Evolution (LTE) is often called “4G”, but many also claim that LTE release 10, also referred to as LTE-Advanced, is the true 4G evolution step, with the first release of LTE (release 8) then being labeled as “3.9G”. Currently, fourth generation wireless access systems using Long Term Evolution (LTE) are being deployed by many operators worldwide in order to offer faster access with lower latency and more efficiency than 3G/3.5G [1-2]. Orthogonal Frequency Division Multiple Access (OFDMA) and Single Carrier Frequency Division Multiple Access (SC-FDMA) are modified versions of the OFDM and SC/FDE schemes. The OFDMA and SC-FDMA are used in LTE cellular systems for downlink and uplink transmission and also making the mobile terminal power-efficient [3].

The present study has been made for performance analysis of a CRC/Convolutional encoded SC- FDMA wireless communication system with implementation of MMSE-SIC and ZF- SIC channel equalization schemes. In 2013, Sadique and Ullah made performance evaluation study for a STBC transmission scheme based Turbo encoded SC-FDMA wireless communication system on encrypted data transmission. In such simulation work, three linear signal detection techniques (Equalizers) such as Minimum Mean Square Error (MMSE), Zero Forcing (ZF) and Q-less QR Decomposition were used [4].

#### **MATHEMATICAL MODEL**

##### **3-slot Orthogonal Space-time Block Coding**

The 3-slot orthogonal space-time block coding is an orthogonal STBC (Q-STBC) coding scheme and such scheme has been found to have superior performance with achievement of rate-one full-diversity and reduced decoding complexity. In 3-slot STBC scheme, three consecutive frequency domain data symbols  $X_1$ ,  $X_2$  and  $X_3$  and their modified form are entered into a  $2 \times 3$  data matrix,  $X$  of Equation(1)

$$\mathbf{X} = \begin{bmatrix} x_1 & x_2 & x_3 \\ \frac{x_1^* + 2e^{j\frac{2\pi}{5}} x_2^* + 2e^{-j\frac{2\pi}{5}} x_3^*}{3} & \frac{-2e^{j\frac{2\pi}{5}} x_1^* + e^{-j\frac{\pi}{5}} x_2^* + 2x_3^*}{3} & \frac{-2e^{-j\frac{2\pi}{5}} x_1^* + 2x_2^* + e^{j\frac{\pi}{5}} x_3^*}{3} \end{bmatrix} \quad (1)$$

Generally, the symbols  $x_1$ ,  $x_2$  and  $x_3$  and their linear combinations (second rows of data matrix  $\mathbf{X}$ ) are processed separately in three blocks of 64 symbols each as 64-point FFT operation is done on digitally modulated symbols prior to 3-slot orthogonal space-time block encoding scheme. The data symbols and their Q-STBC encoded symbols are mapped onto 64 consecutive subcarriers and converted into time domain signal by 1024-point IFFT to transmit from the first and second antenna respectively [5]. With channel equalization technique, the transmitted frequency domain signals from each of the two antennas are detected. The detected signals are processed furthermore to recover data symbols.

**Channel Equalization**

With channel equalization technique, the transmit signal  $\mathbf{s}$  is detected from the received signal  $\mathbf{y}$  under the knowledge of estimated channel state information (CSI) of  $2 \times 2$  channel matrix  $\mathbf{H}$  and the statistical properties of zero-mean complex Gaussian random noise  $\mathbf{n}$ . The received signal  $\mathbf{y}$  can be written as:

$$\mathbf{y} = \mathbf{H}\mathbf{s} + \mathbf{n} \quad (2)$$

In ZF-SIC channel equalization scheme,  $\mathbf{H}$  undergoes QR factorization as

$$\mathbf{H} = \mathbf{Q}\mathbf{R} = \mathbf{Q} \begin{bmatrix} r_{1,1} & r_{1,2} \\ \mathbf{0} & r_{2,2} \end{bmatrix} \quad (3)$$

where,  $\mathbf{Q}$  and  $\mathbf{R}$  are the unitary and upper triangular matrix respectively. Equation (2) can be rewritten on multiplying by  $\mathbf{Q}^H$  as

$$\begin{aligned} \mathbf{x} &= \mathbf{Q}^H \mathbf{y} \\ &= \mathbf{R}\mathbf{s} + \mathbf{Q}^H \mathbf{n} \end{aligned} \quad (4)$$

where,  $\mathbf{Q}^H \mathbf{n}$  is a zero-mean complex Gaussian random vector. Since  $\mathbf{Q}^H \mathbf{n}$  and  $\mathbf{n}$  have the same statistical properties,  $\mathbf{Q}^H \mathbf{n}$  can be used to denote  $\mathbf{n}$ . We get Equation (4) as

$$\mathbf{x} = \mathbf{R}\mathbf{s} + \mathbf{n}$$

⇓

$$\begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} r_{1,1} & r_{1,2} \\ 0 & r_{2,2} \end{bmatrix} \begin{bmatrix} s_1 \\ s_2 \end{bmatrix} + \begin{bmatrix} n_1 \\ n_2 \end{bmatrix} \quad (5)$$

the detected desired signal  $\hat{\mathbf{s}}$  from the transmitting antennas can be written on neglecting noise term from Equation (5) as

$$\hat{\mathbf{s}}_1 = \frac{\{x_1 - r_{1,2}(x_2/r_{2,2})\}}{r_{1,1}} \quad \text{and} \quad \hat{\mathbf{s}}_2 = \frac{x_2}{r_{2,2}} \quad (6)$$

In MMSE-SIC channel equalization scheme, the received signal, channel matrix and noise are extended as

$$\mathbf{H}_{ex} = \begin{bmatrix} \mathbf{H}^T \sqrt{\frac{N_0}{E_s}} \mathbf{I} \end{bmatrix}^T, \mathbf{y}_{ex} = [\mathbf{y}^T \mathbf{0}^T] \text{ and } \mathbf{n}_{ex} = \begin{bmatrix} \mathbf{n}^T - \sqrt{\frac{N_0}{E_s}} \mathbf{s}^T \end{bmatrix}^T \quad (7)$$

where,  $\frac{N_0}{E_s} = 1/(\text{signal to noise ratio, SNR})$ . On QR factorization of extended channel matrix,  $H_{ex}$ , we get

$$H_{ex} = Q_{ex} R_{ex} \tag{8}$$

Where,  $Q_{ex}$  and  $R_{ex}$  represent a unitary matrix and an upper triangular matrix respectively.

In Equation (4), we assume that  $y$ ,  $H$ ,  $n$ ,  $Q$  and  $R$  are replaced by  $y_{ex}, H_{ex}, n_{ex}, Q_{ex}$  and  $R_{ex}$  respectively and correspondingly the resulting system takes the following form

$$\begin{aligned} x_{ex} &= Q_{ex}^H y_{ex} \\ &= R_{ex} s + Q_{ex}^H n_{ex} \end{aligned} \tag{9}$$

Neglecting  $Q_{ex}^H n_{ex}$  term, the detected desired signal  $\hat{s}$  from the transmitting antennas can be written from Equation (9) as [6],

$$\hat{s}_1 = \frac{\{x_{ex1} - r_{ex1,2}(x_{ex2}/r_{ex2,2})\}}{r_{ex1,1}} \quad \text{and} \quad \hat{s}_2 = \frac{x_{ex2}}{r_{ex2,2}} \tag{10}$$

**SYSTEM MODEL**

A simulated 3 slot STBC scheme implemented SC-FDMA wireless communication system as depicted in Figure 1 utilizes ZF-SIC and MMSE-SIC linear channel equalization schemes. In such a simulated communication system, an audio signal with double channel data is extracted from a video file using Video converter. Using MATLAB, a single channel audio signal is eliminated and processed for A/D Conversion using PCM encoding. The A/D converted binary data are scrambled, channel coded, interleaved and digitally modulated using three types of digital modulations (DQPSK, QPSK and QAM) [7-8]. The digitally modulated data symbols are fed into 64-point FFT block and subsequently sent to Q-STBC encoding section. The quasi orthogonal STBC encoded complex data symbols are processed for subcarrier mapping, 1024-point IFFT block and Cyclic Prefixing operations and eventually transmitted from each of the two transmitting antenna. At the receiver end, reverse operation is done viz. after detecting the transmitted signals using channel equalization schemes. The signals are then processed with removal of cyclic prefixing. The signals are then fed into 1024-point FFT, subcarrier demapping and Q-STBC decoding sections. The decoded data are digitally demodulated, de-interleaved, channel decoded and descrambled and subsequently used for D/A conversion using PCM decoding scheme for retrieving transmitted audio signal.

**RESULTS AND DISCUSSION**

In our analysis, we have assumed that the MIMO channel state information (CSI) is available at the transmitter side and the fading process is constant during each transmitted signal. Figure 2 through Figure 5 depict the bit-error rate performance of the FEC encoded SC-FDMA system for different channel coding, channel equalization and digital modulation schemes. The simulation parameters used in our study are presented in Table 1.

In Figure 2, it is noticeable that the convolutionally encoded SC-FDMA system with DQPSK and MMSE-SIC shows almost flat BER performance at low SNR value area. In such a case, performance improvement can be observed with increase in SNR values. From this Figure, we can see that the system outperforms with QAM digital modulation. At 1% bit error rate, the convolutionally encoded SC-FDMA system with QAM is superior by 1.2 dB and 3.8 dB respectively as compared with QPSK and DQPSK. In Figure 3, it is seen that the BER performance becomes poorer in case of convolutionally encoded SC-FDMA system with DQPSK and ZF-SIC. The BER performances in three different digital modulations are quite distinct. At a typically assumed SNR value of 5dB, the estimated BER values are 0.0329 and 0.3451 for QAM and DQPSK which implies system performance improvement by 10.21 dB.

In Figure 4, it is found that the CRC encoded SC-FDMA system with MMSE-SIC provides almost identical performance over a significant SNR value area for DQPSK and QPSK digital modulations. At 1% bit error rate, the CRC encoded SC-FDMA system with QAM is superior by 4 dB as compared to QPSK/ DQPSK. The system outperforms in QAM digital modulation. In Figure 5, the estimated BERs are 0.0388 and 0.1464 in case of QAM and DQPSK at SNR value of 5dB for the CRC encoded SC-FDMA system with ZF-SIC which is indicative of system performance improvement by 5.77 dB. At 1% bit error rate, the CRC encoded SC-FDMA system with QAM is superior by 1.2 dB and 4.1 dB respectively as compared with QPSK and DQPSK under implemented ZF-SIC channel equalization

scheme. In Figure 6, the transmitted and retrieved audio signals are presented graphically. The estimated bit error rate are 0.1551, 0.1024, 0.0572, 0.0237 and 0.0008 in case of 0 2 4 6 and 10 dB SNR values. In Figure 7, the transmitted audio signals have been compared with retrieved audio signal at 10 dB SNR value in spectral form. The graphical illustration presented in this Figure confirms satisfactory retrieval of transmitted audio signal in such a system.

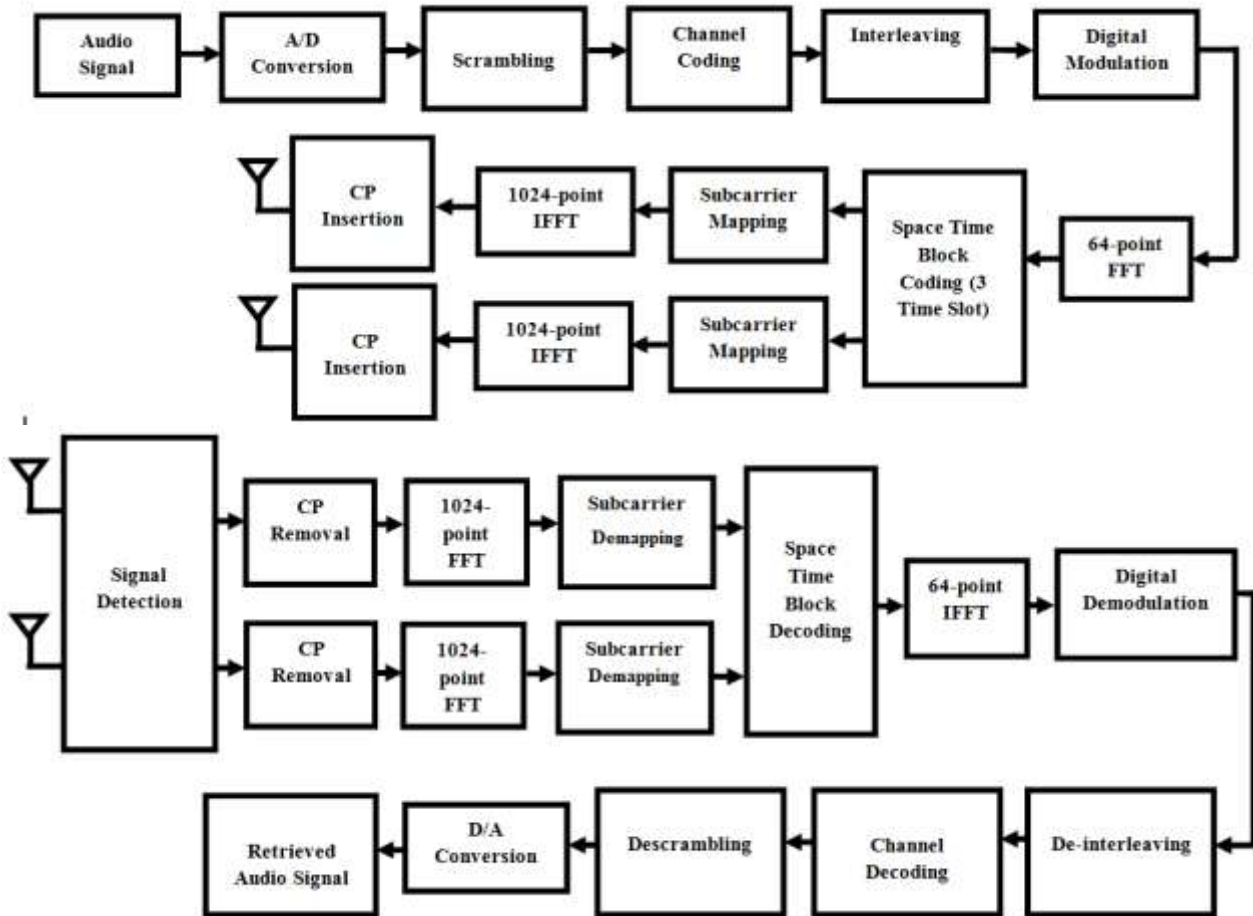


Fig-1: Block diagram of a scrambled audio frequency transmission scheme based SC-FDMA wireless communication system.

Table-1: Summary of the simulated model parameters

|  |                                 |
|--|---------------------------------|
| No. of Analog Sample of Segmented Audio Signal | 3996                            |
| No. of bits used for segmented audio signal    | 63936                           |
| Sampling frequency of audio signal             | 48 KHz                          |
| Antenna Configuration                          | 2-by-2                          |
| Channel Coding /Decoding                       | CRC and 1/2-rated Convolutional |
| Modulation                                     | DQPSK, QPSK and QAM             |
| No. of OFDM sub-carriers                       | 1024                            |
| Signal Detection Scheme                        | MMSE-SIC and ZF-SIC             |
| CP length                                      | 103 symbols                     |
| Channel  | AWGN and Rayleigh               |
| Signal to noise ratio, SNR                     | 0 to10 dB                       |

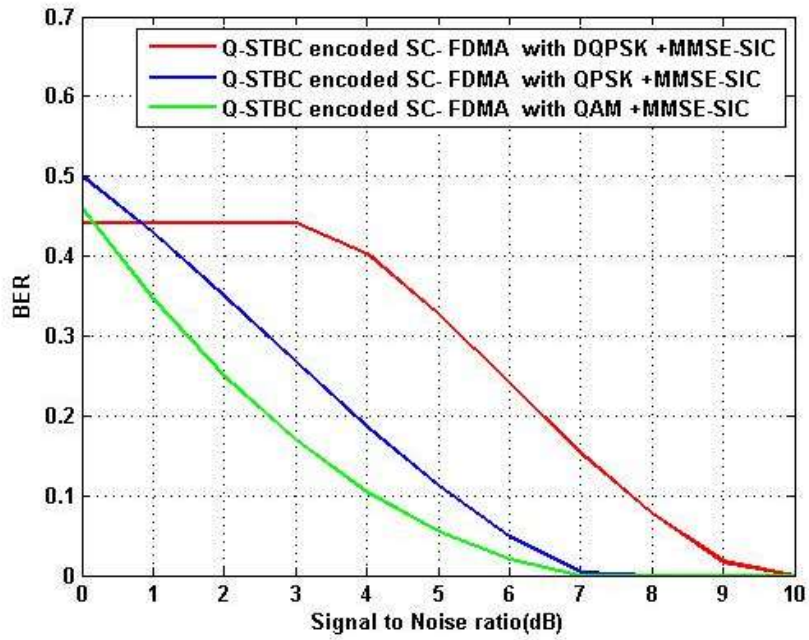


Fig-2: Performance comparison of a Convolutiionally encoded SC-FDMA based system with MMSE-SIC based signal detection scheme and various digital modulation schemes

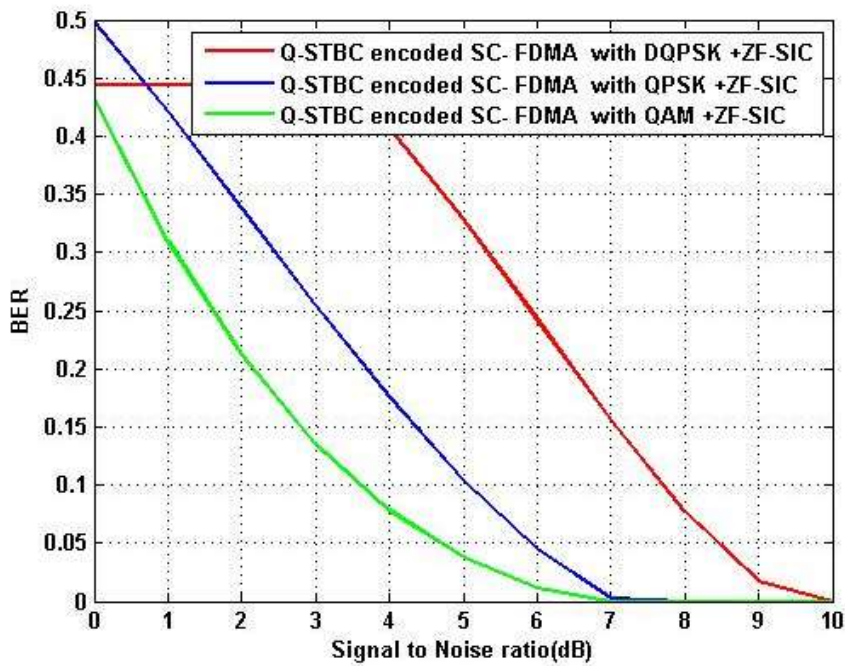


Fig-3: Performance comparison of a convolutiionally encoded SC-FDMA based system with ZF-SIC based signal detection scheme and various digital modulation schemes

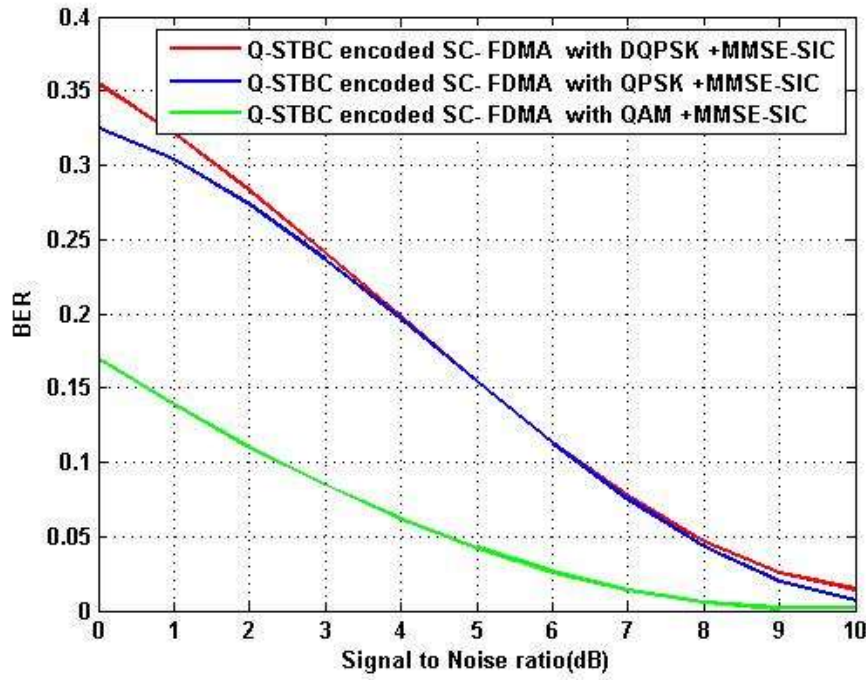


Fig- 4: Performance comparison of a CRC encoded SC-FDMA based system with MMSE-SIC based signal detection scheme and various digital modulation schemes

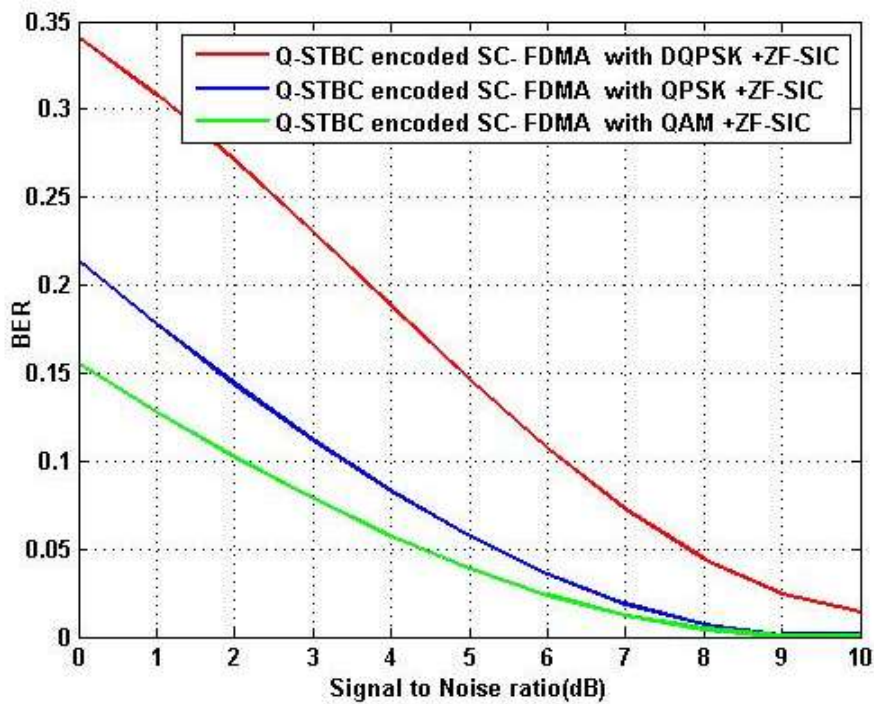
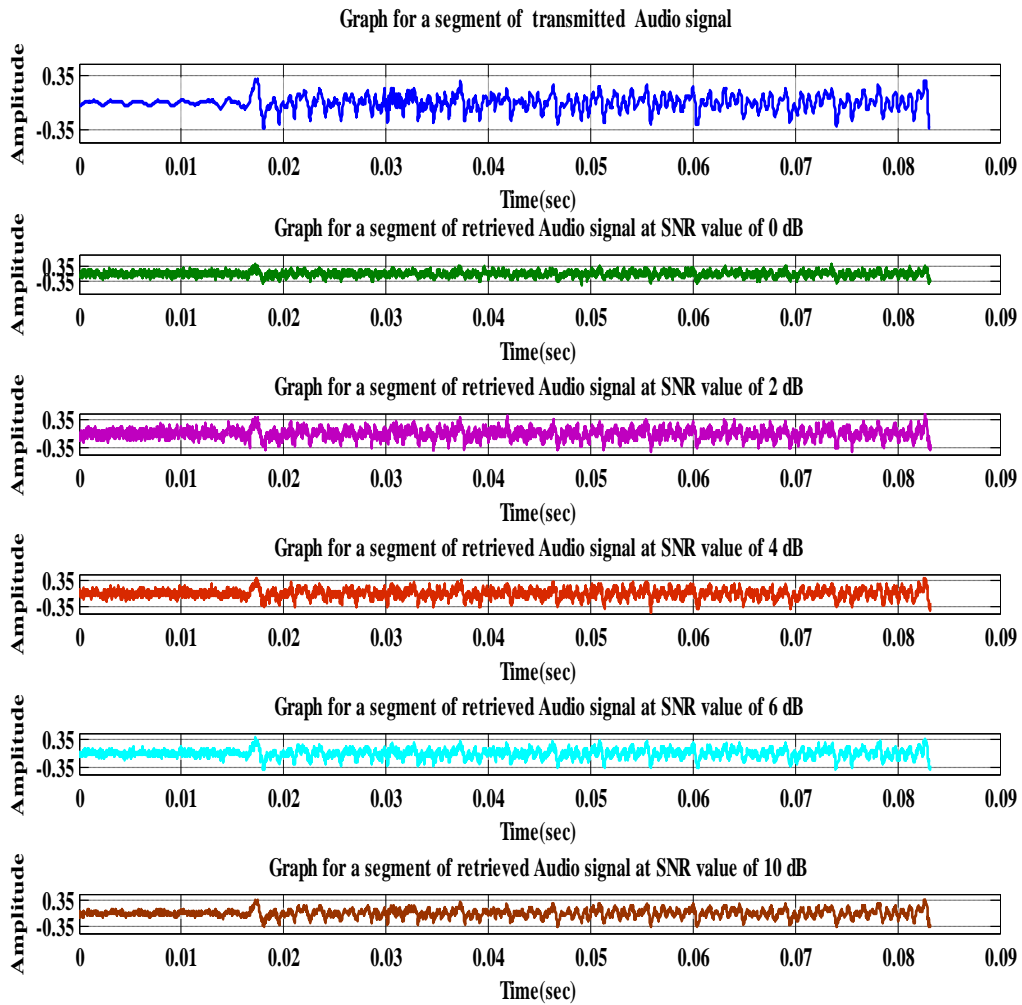
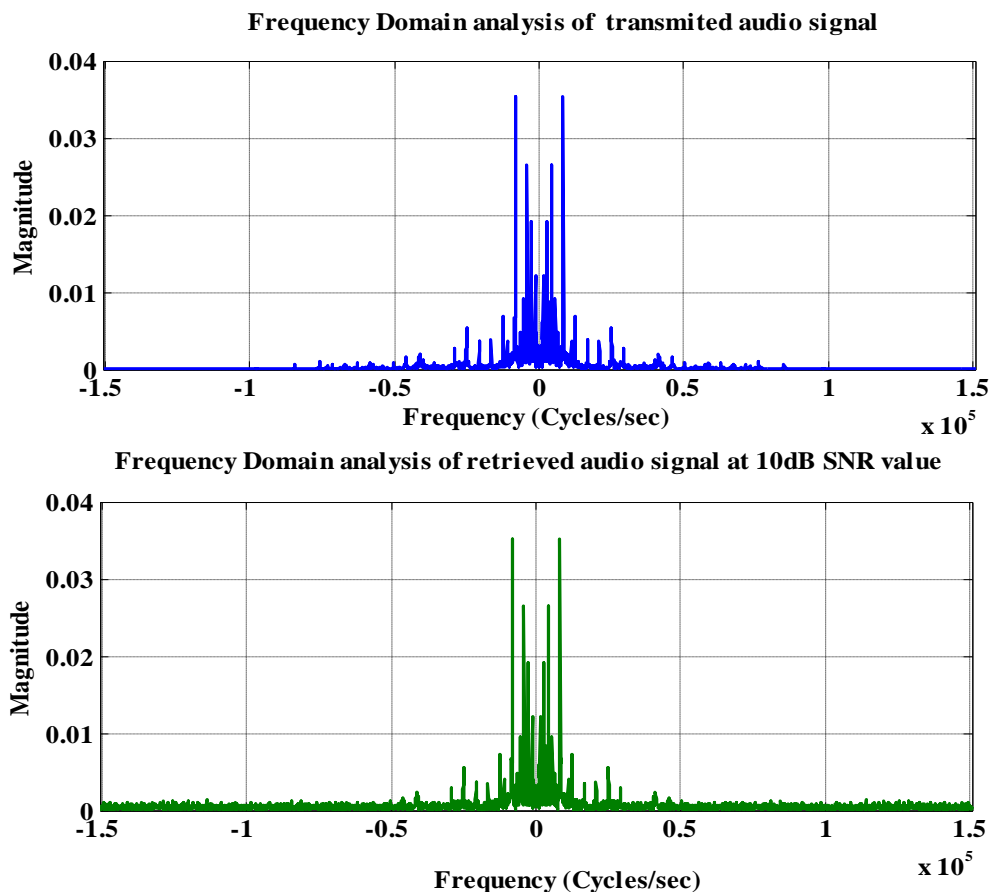


Fig-5: Performance comparison of a CRC encoded SC-FDMA based system with ZF-SIC based signal detection scheme and various digital modulation schemes



**Fig-6: Transmitted and Retrieved voice frequency signals with implementation of ZF-SIC signal detection and QAM digital modulation schemes in a CRC encoded SC-FDMA wireless communication system**



**Fig-7: Comparison of amplitude spectrum of Transmitted and Retrieved audio frequency signals with implementation of ZF-SIC signal detection and QAM digital modulation schemes in a CRC encoded SC-FDMA wireless communication system**

## CONCLUSION

Within this paper, the BER performance of a 3-slot STBC based  $2 \times 2$  spatially multiplexed SC-FDMA wireless communication system combined with transmit diversity and receive diversity in Rayleigh fading channel has been analyzed. We have shown the impact of various channel coding and channel equalization techniques in BER performance evaluative study of the SC-FDMA system on retrieval of transmitted audio signal. A range of satisfactory BERs at different low SNR values under Q-STBC, QAM, CRC and ZF-SIC schemes highlights the improved link reliability and performance enhancement of the FEC encoded MIMO SC-FDMA Wireless Communication System.

## REFERENCES

1. Erik D, Parkvall S, Sköld J; 4G LTE/LTE-Advanced for Mobile Broadband, Elsevier publisher Ltd, United Kingdom, 2011.
2. Nakamura T, Satoshi N, Benjebbour A, Kishiyama Y, Tang H, Shen X, Yang N, Li N; Trends in Small Cell Enhancements in LTE Advanced, IEEE Communications Magazine, 2013; 98-105.
3. Hyung GM, and David JG; Single carrier FDMA- a new air interface for Long term evolution, John Wiley and Sons, Ltd, Publication, UK. 2008.
4. Sadique JJ, Enayet SU; Encrypted Data Transmission in STBC Transmission Scheme Based Turbo Encoded SC-FDMA Wireless Communication System, International Journal of Advanced Science and Technology, 2013; 50:1-10
5. Lei Z, Yuen C, Francois PS; Quasi-Orthogonal Space-Time Block Codes for Two Transmit Antennas and Three Time Slots, IEEE Transactions on Wireless Communications, 2011; 10(6):1983-1991.
6. Lin B, Choi J; Low Complexity MIMO Detection, Springer Science and Business Media, LLC, New York, USA, 2012.
7. Lee BG, Kim SC; Scrambling Techniques for Digital Transmission, Springer Verlag, 1994.
8. Theodore S, Rappaport T; Wireless communications: Principles and Practices, Second Edition, Prentice Hall Inc., New Jersey, USA, 2008.